

# **Analog Electronics**

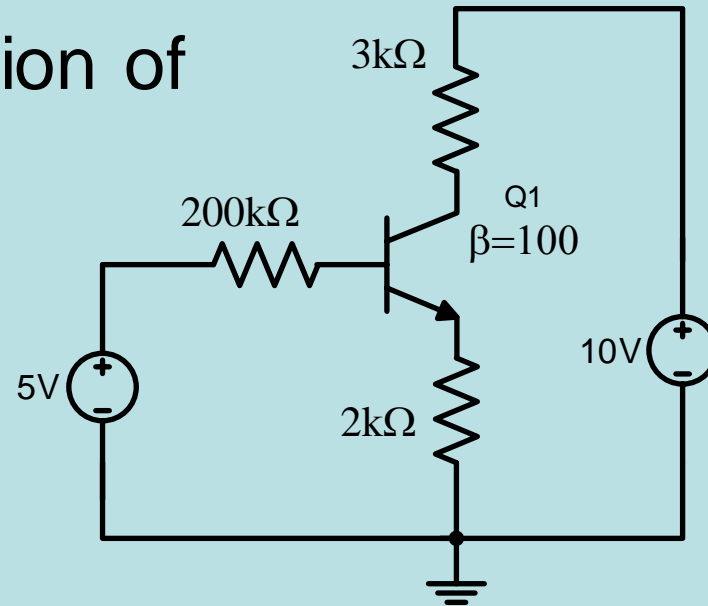
## **ENEE236**

Instructor: Nasser Ismail

**L8- DC Biasing - BJTs**

# Example

- Assume  $V_{CE(sat)}=0.2\text{ V}$
- Find mode of operation of Q1 ?



# Determine Mode of Operation of BJT?

- Solution:
- 1) Since BE junction is forward biased ==> Q1 can be either in Active (Linear) or Saturation mode
- Assume it is in Active Mode

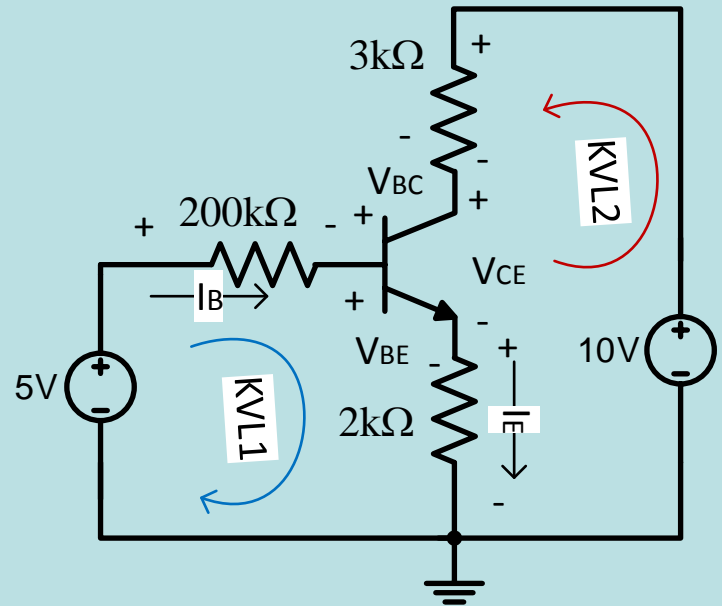
$$5 = 200 \text{ k}\Omega \cdot I_B + V_{BE} + 2 \text{ k}\Omega \cdot I_E$$

But,  $I_E = (1 + \beta)I_B$

$$\text{Solve for } I_B = \frac{5 - V_{BE}}{200 \text{ k}\Omega + (1 + \beta) \cdot 2 \text{ k}\Omega}$$

$$I_B = \frac{5 - 0.7}{200 \text{ k}\Omega + (1 + 100) \cdot 2 \text{ k}\Omega}$$

$$= \frac{4.3 \text{ V}}{402 \text{ k}\Omega} = 10.7 \text{ } \mu\text{A}$$



$$I_C = \beta I_B$$

$$= (100) \cdot (10.7 \mu\text{A})$$

$$= 1.07 \text{ mA}$$

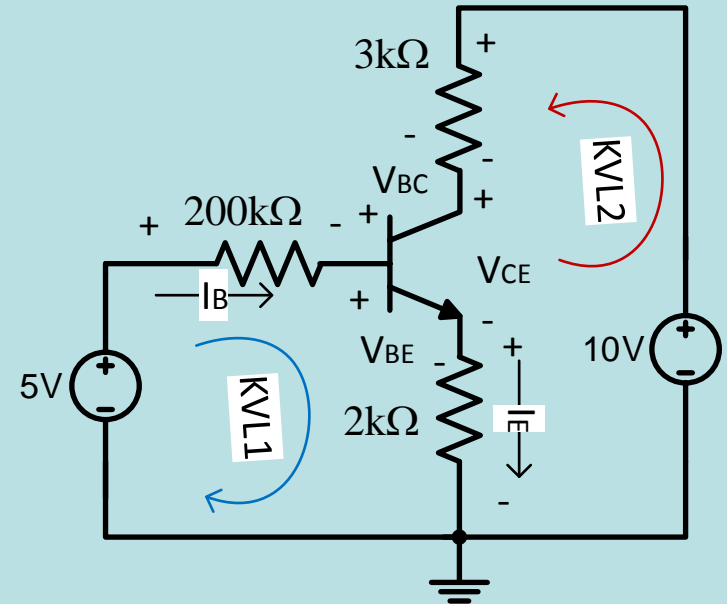
$$I_E = (\beta + 1) I_B$$

$$= 1.0807 \text{ mA}$$

Now we find  $V_{CE}$  from output circuit

$$10 - I_C \cdot 3 \text{ k}\Omega - I_E \cdot 2 \text{ k}\Omega = V_{CE}$$

$$\Rightarrow V_{CE} = 4.63 \text{ V} > V_{CE(\text{sat})}$$



$\therefore$  Q1 is in active mode and the assumption is true  
 we can also verify that the BC junction is reverse  
 biased which is required so that the BJT operates  
 in active mode

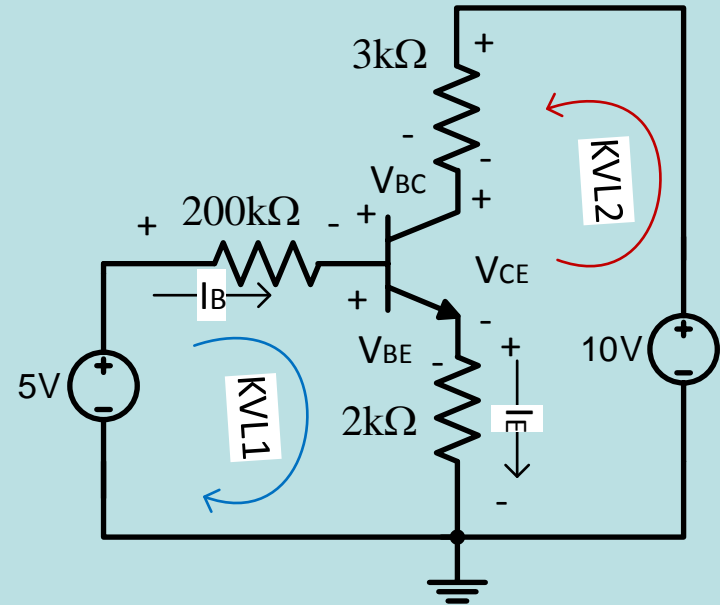
$$10 - I_C \cdot 3 \text{ k}\Omega - I_E \cdot 2 \text{ k}\Omega = V_{CE}$$

$$\Rightarrow V_{CE} = V_{CB} - V_{EB}$$

$$\Rightarrow V_{CB} = V_{CE} - V_{BE} = 4.63 - 0.7 = 3.93 \text{ V}$$

$$\therefore V_{BC} = -V_{CB} = -3.33 \text{ V}$$

BC junction is reverse biased



- Solution:
- 1) Since BE junction is forward biased  $\implies$  Q1 can be either in Active (Linear) or Saturation mode
- Assume it is in saturation mode:

$$10 - I_{C(\text{sat})} \cdot 3\text{k}\Omega - I_{E(\text{sat})} \cdot 2\text{k}\Omega = V_{CE(\text{Sat})}$$

assume  $I_{E(\text{sat})} = I_{C(\text{sat})}$

$$\therefore I_{C(\text{sat})} = \frac{10 - 0.2}{5\text{k}\Omega} = 1.96 \text{ mA}$$

$$I_{B(\text{min})} = \frac{I_{C(\text{sat})}}{\beta} = 19.6 \text{ }\mu\text{A}$$

Now we find the actual value of  $I_B$

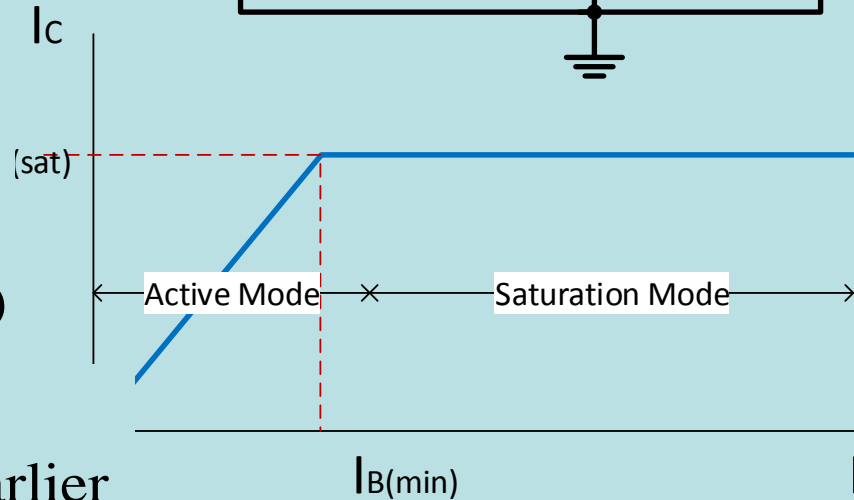
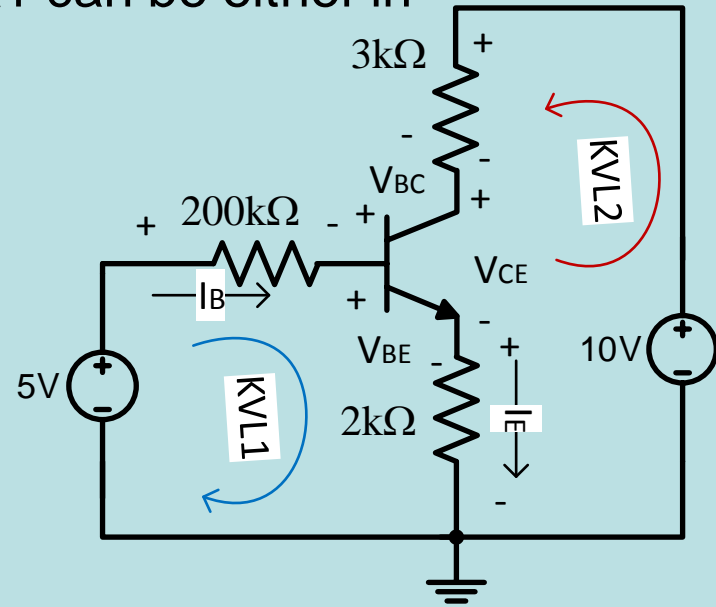
$$I_{B(\text{actual})} = 10.7 \text{ }\mu\text{A} \text{ (it was found previously)}$$

since

$$I_{B(\text{actual})} < I_{B(\text{sat})} \implies \text{the assumption made earlier}$$

that BJT in saturation mode is wrong ,

and actually it is in active mode

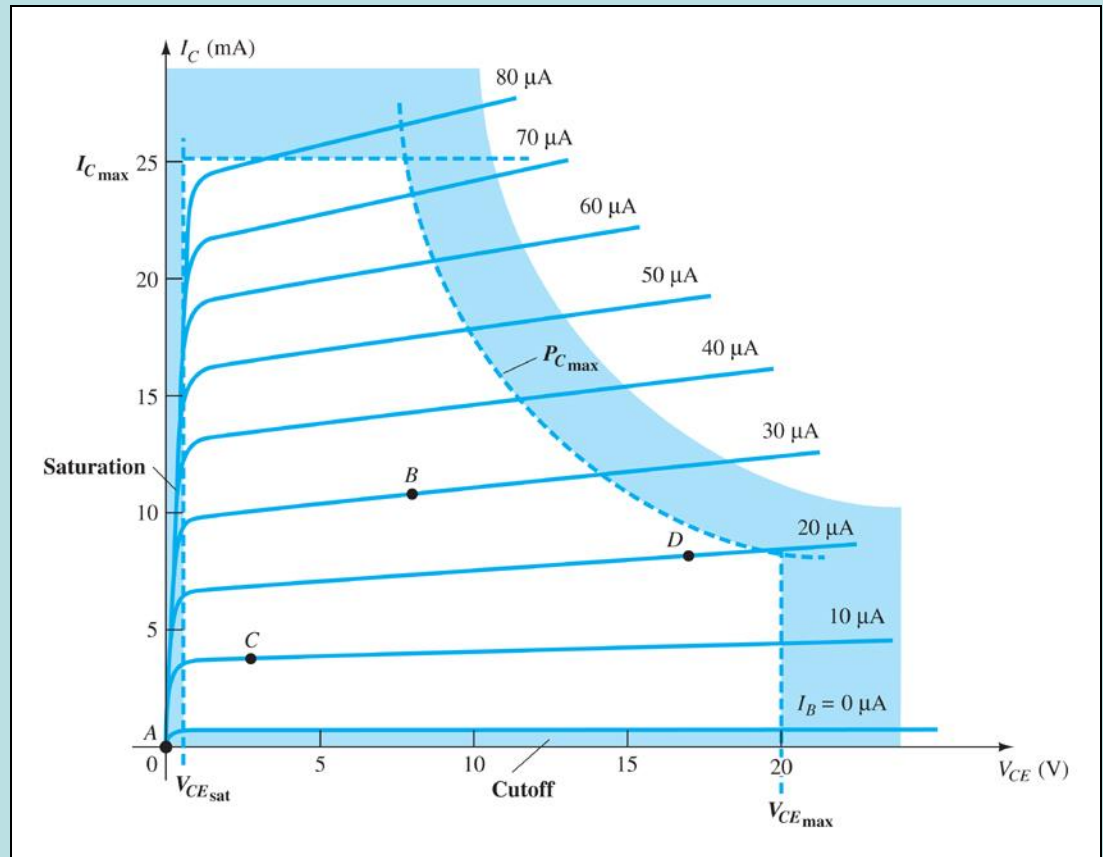


# Biasing

**Biasing:** Applying DC voltages to a transistor in order to establish fixed level of voltage and current, for Amplifier mode, the resulting dc voltage and current establish the operation point to turn it on so that it can amplify AC signals.

# Operating Point

The DC input establishes an operating or *quiescent point* called the **Q-point**.





# The Three Operating Regions

## Active or Linear Region Operation

- Base–Emitter junction is forward biased
- Base–Collector junction is reverse biased

## Cutoff Region Operation

- Base–Emitter junction is reverse biased

## Saturation Region Operation

- Base–Emitter junction is forward biased
- Base–Collector junction is forward biased

# DC Biasing Circuits

Fixed-bias circuit

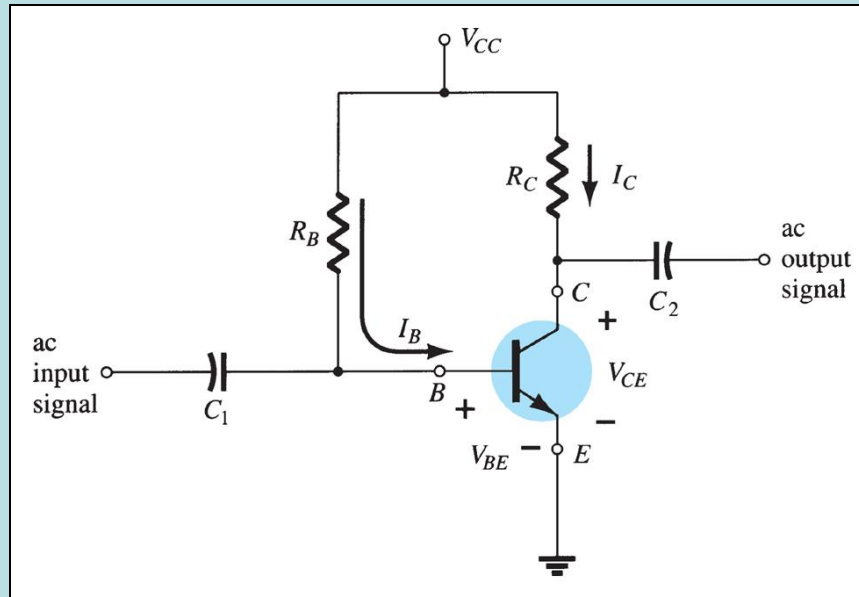
Emitter-stabilized bias circuit

Collector-emitter loop

Voltage divider bias circuit

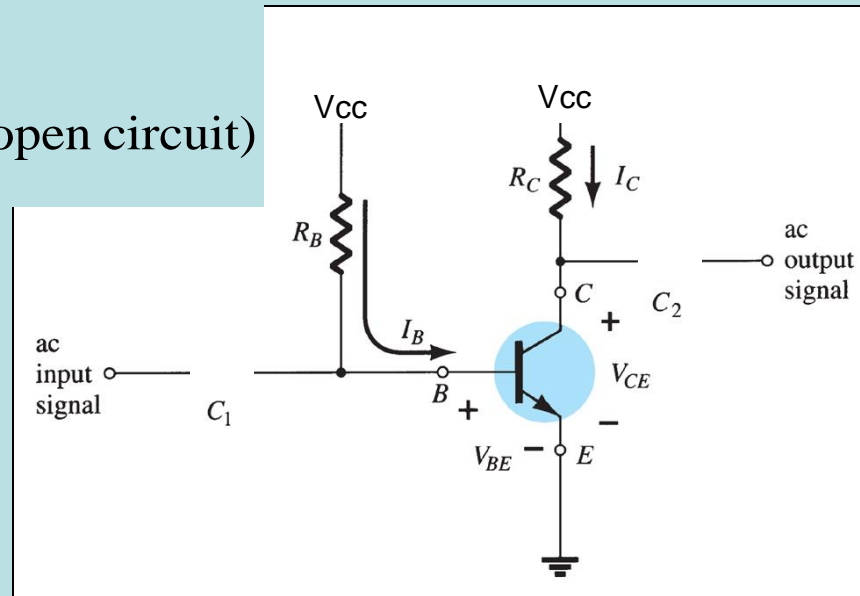
DC bias with voltage feedback

# 1) Fixed Bias Configuration



DC equivalent circuit

$$f = 0 \Rightarrow X_c = \frac{1}{2\pi f C} \cong \infty \text{ (open circuit)}$$



# The Base-Emitter Loop

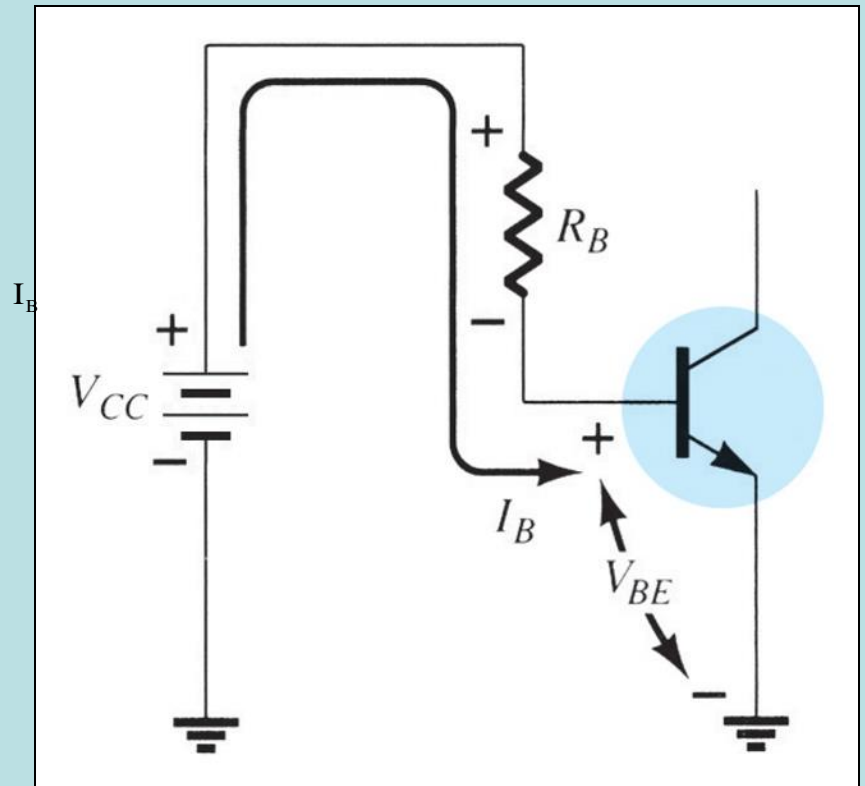
From Kirchhoff's voltage law for Input:

$$+V_{CC} - I_B R_B - V_{BE} = 0$$

Solving for base current:

$$I_B = \frac{V_{CC} - V_{BE}}{R_B}$$

Choosing  $R_B$  will establish the required level of  $I_B$



# Collector-Emitter Loop

Collector current:

$$I_C = \beta I_B$$

From Kirchhoff's voltage law:

$$V_{CE} = V_{CC} - I_C R_C$$

$$V_{CE} = V_C - V_E$$

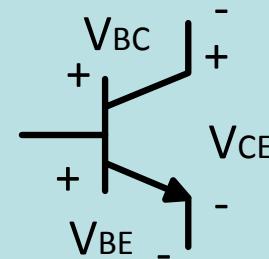
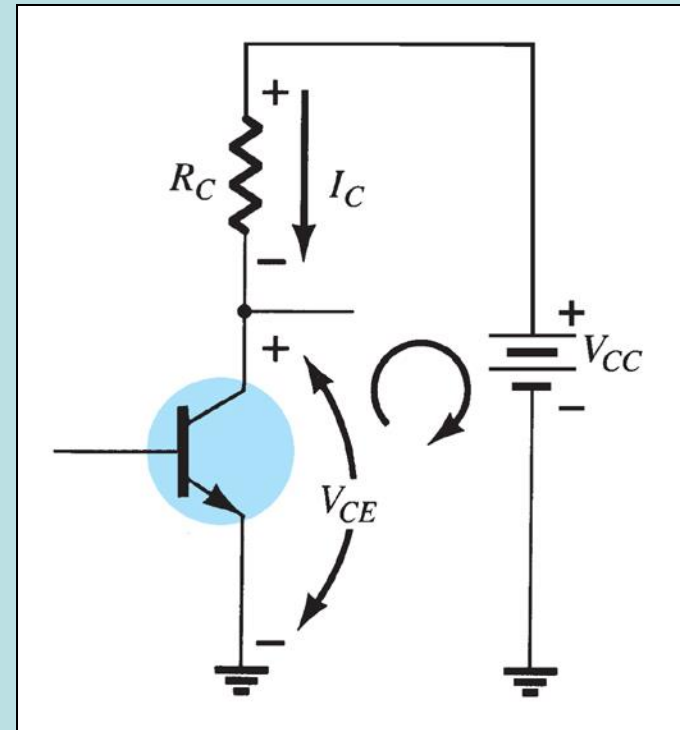
Since  $V_E = 0$

$$\therefore V_{CE} = V_C$$

Also

$$V_{BE} = V_B -$$

$$\therefore V_{BE} = V_B$$



$$V_{BE} - V_{CE} - V_{BC} = 0$$

$$\therefore V_{BC} = V_{BE} - V_{CE}$$

# Saturation

When the transistor is operating in **saturation**, current through the transistor is at its *maximum* possible value.

$$I_{Csat} = \frac{V_{CC}}{R_C}$$

$$V_{CE} \cong 0 \text{ V}$$

# Ch.4 Summary

## Load Line Analysis

The load line end points are:

**$I_{Csat}$**

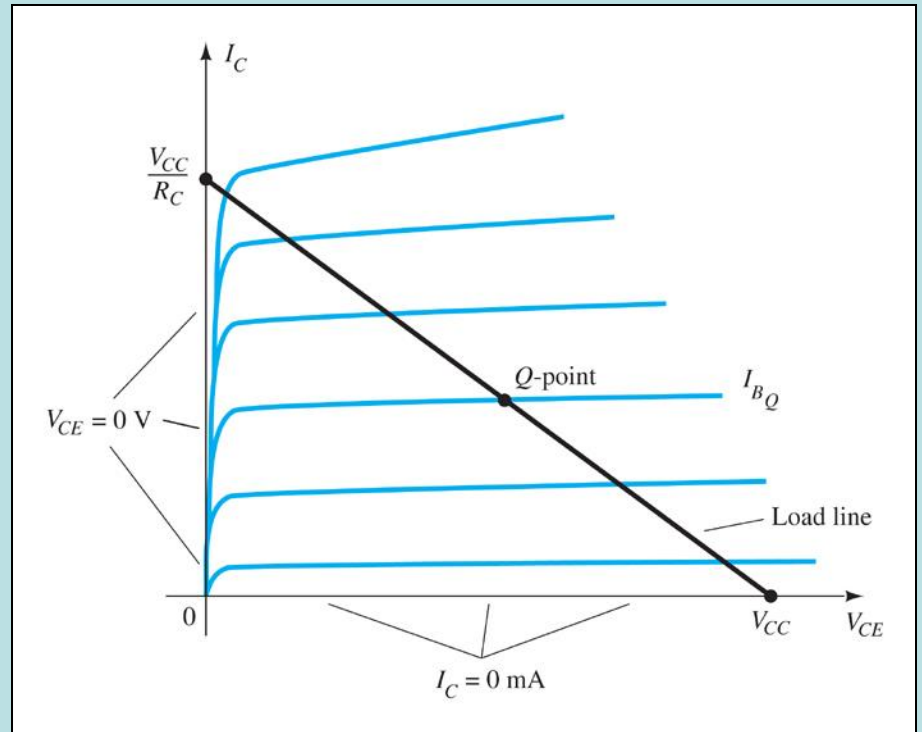
$$I_C = V_{CC} / R_C$$

$$V_{CE} = 0 \text{ V}$$

**$V_{CEcutoff}$**

$$V_{CE} = V_{CC}$$

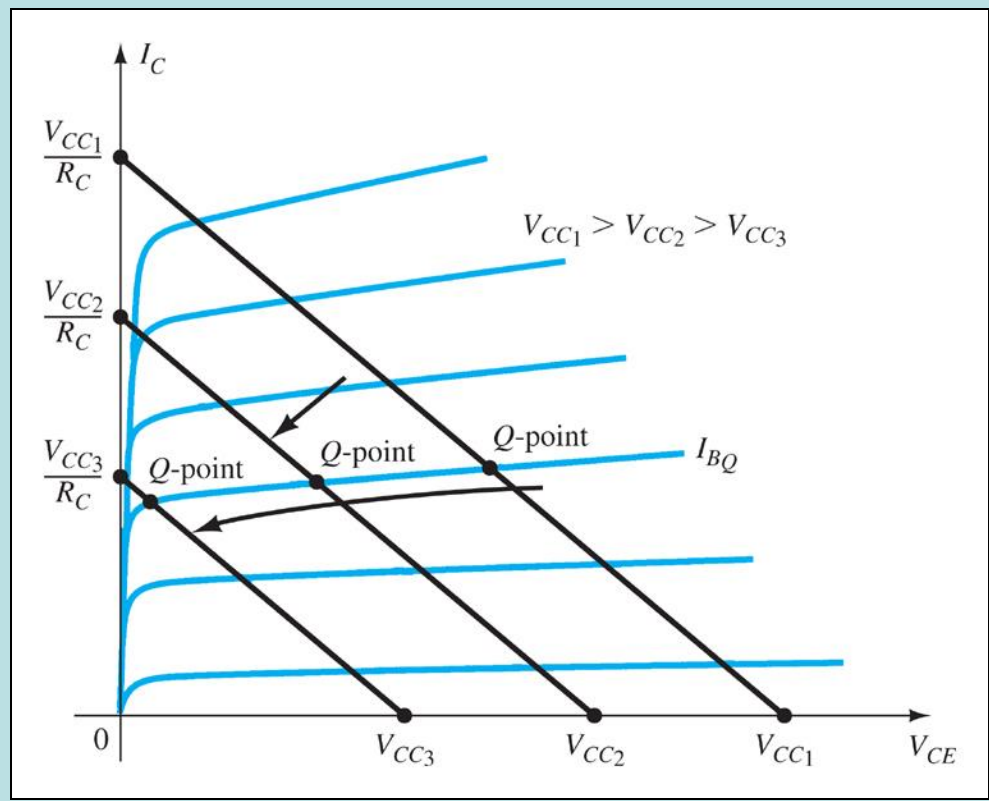
$$I_C = 0 \text{ mA}$$



The Q-point is the operating point where the value of  $R_B$  sets the value of  $I_B$  that controls the values of  $V_{CE}$  and  $I_C$ .

# Ch.4 Summary

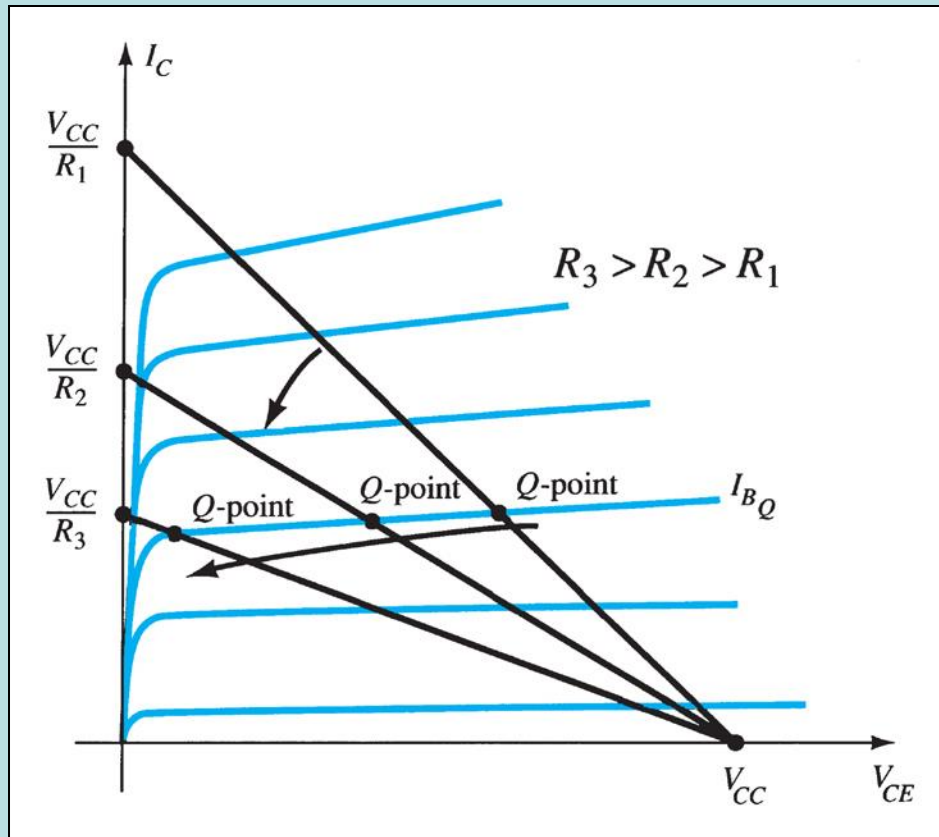
## The Effect of $V_{CC}$ on the Q-Point



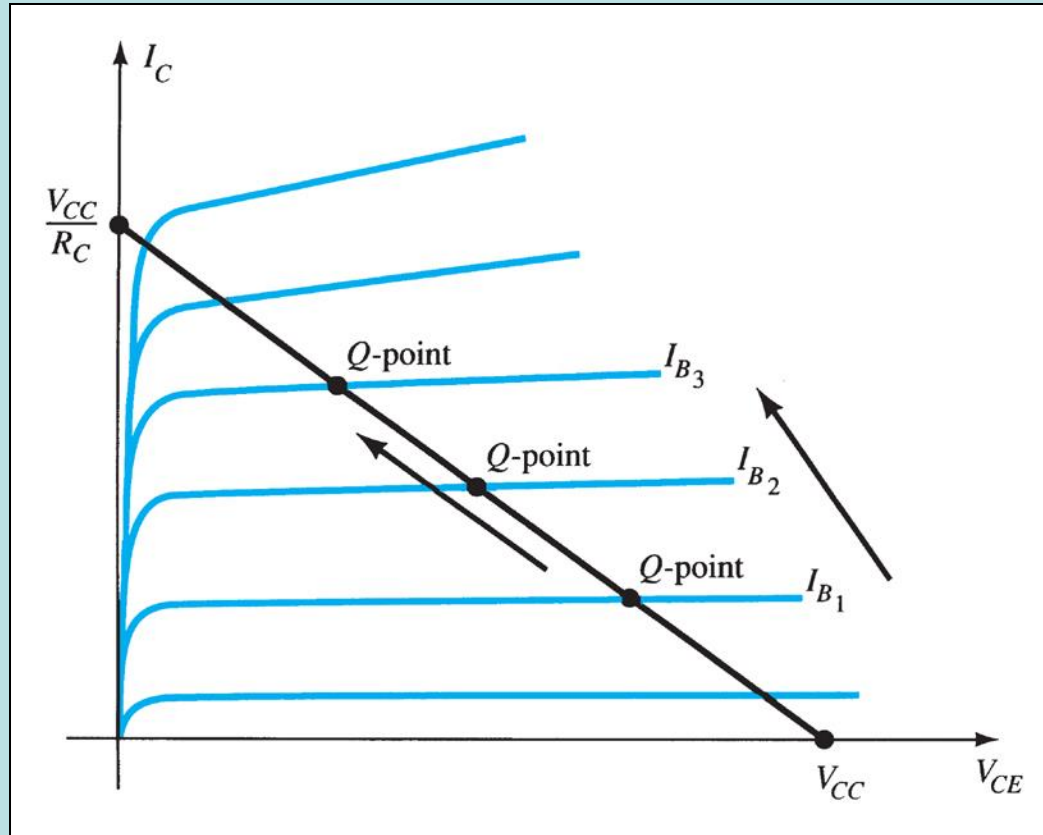


# Ch.4 Summary

## The Effect of $R_C$ on the Q-Point



# The Effect of $I_B$ on the Q-Point



# Design: Fixed bias

Assume  $V_{CC} = 10V$ ,  $\beta_{\text{nominal}} = 100$ ,  $\beta_{\text{min}} = 50$ ,  $\beta_{\text{max}} = 150$

Design for Q-point :  $V_{CEQ} = 5V$ ,  $I_{CQ} = 1mA$

*Solution*

$$I_{BQ} = \frac{I_{CQ}}{\beta_{\text{nominal}}} = \frac{1 \text{ mA}}{100} = 10 \mu\text{A}$$

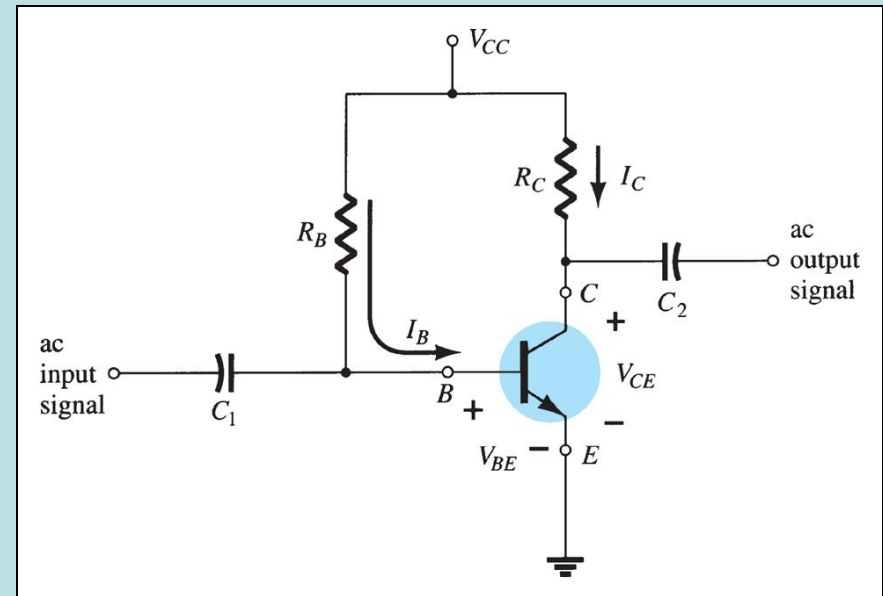
$$I_B = \frac{V_{CC} - V_{BE}}{R_B} \Rightarrow$$

$$R_B = \frac{V_{CC} - V_{BE}}{I_B} = \frac{10 - 0.7}{10 \mu\text{A}} = 930 \text{ k}\Omega$$

$$V_{CE} = V_{CC} - I_C R_C$$

$$V_{CEQ} = 5 = 10 - I_C R_C$$

$$\therefore R_C = \frac{5}{1 \text{ mA}} = 5 \text{ k}\Omega$$



# Fixed bias Stability

Assume  $V_{CC} = 10V$ ,  $\beta_{\text{nominal}} = 100$ ,  $\beta_{\text{min}} = 50$ ,  $\beta_{\text{max}} = 150$

Design for Q - point :  $V_{CEQ} = 5V$ ,  $I_{CQ} = 1\text{mA}$

*Solution*

If  $\beta = \beta_{\text{min}} = 50$

$$I_B = 10 \mu\text{A}$$

$$I_C = \beta I_B = (50)(10 \mu\text{A}) = 0.5 \text{ mA}$$

$$V_{CE} = V_{CC} - I_C R_C$$

$$V_{CEQ} = 10 - (0.5 \text{ mA})(5 \text{ k}\Omega) = 7.5 \text{ V}$$

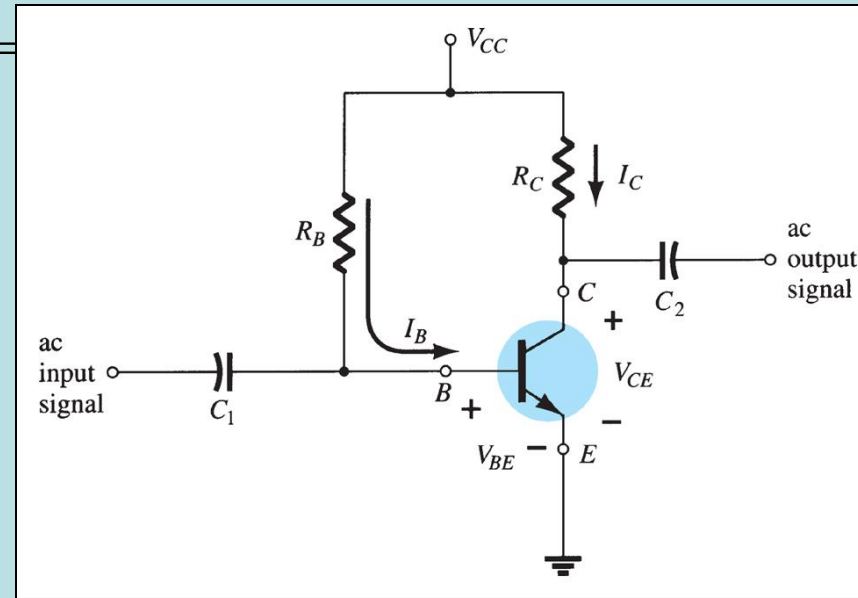
If  $\beta = \beta_{\text{max}} = 150$

$$I_B = 10 \mu\text{A}$$

$$I_C = \beta I_B = (150)(10 \mu\text{A}) = 1.5 \text{ mA}$$

$$V_{CE} = V_{CC} - I_C R_C$$

$$V_{CEQ} = 10 - (1.5 \text{ mA})(5 \text{ k}\Omega) = 2.5 \text{ V}$$



for

$$50 \leq \beta \leq 150$$

$$I_B = 10 \mu\text{A} \text{ fixed}$$

$$0.5 \text{ mA} \leq I_C \leq 1.5 \text{ mA}$$

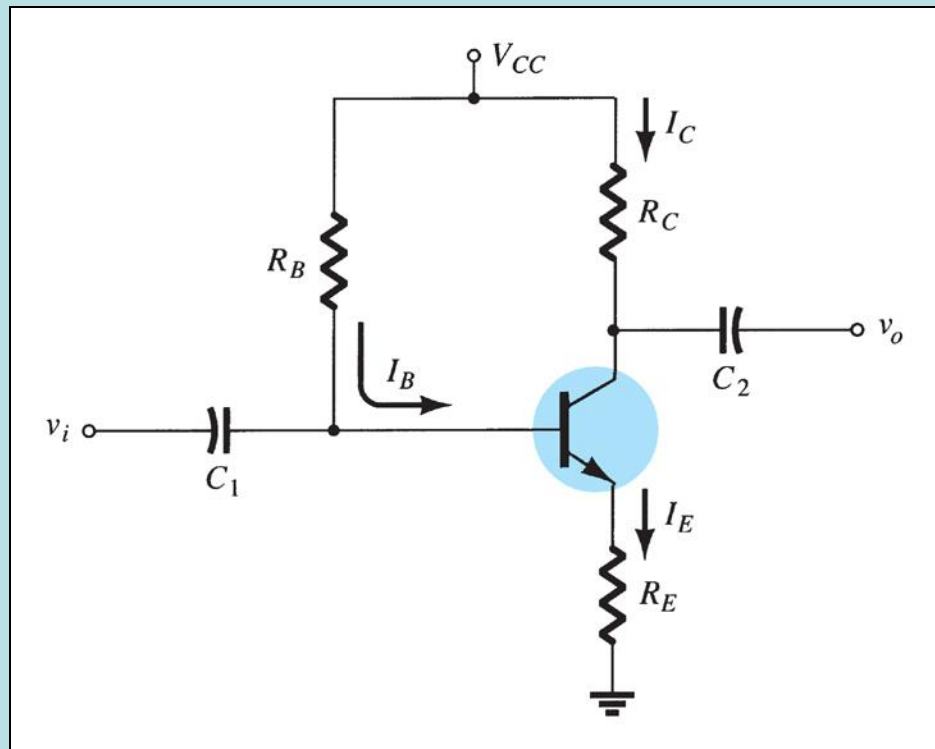
$$7.5 \text{ V} \geq V_{CE} \geq 2.5 \text{ V}$$

$$\therefore \frac{I_{C(\text{max})}}{I_{C(\text{min})}} = \frac{1.5 \text{ mA}}{0.5 \text{ mA}} = 3$$

Not very stable

# Emitter-Stabilized Bias Circuit

Adding a resistor ( $R_E$ ) to the emitter circuit stabilizes the bias circuit.



# Base-Emitter Loop

From Kirchhoff's voltage law:

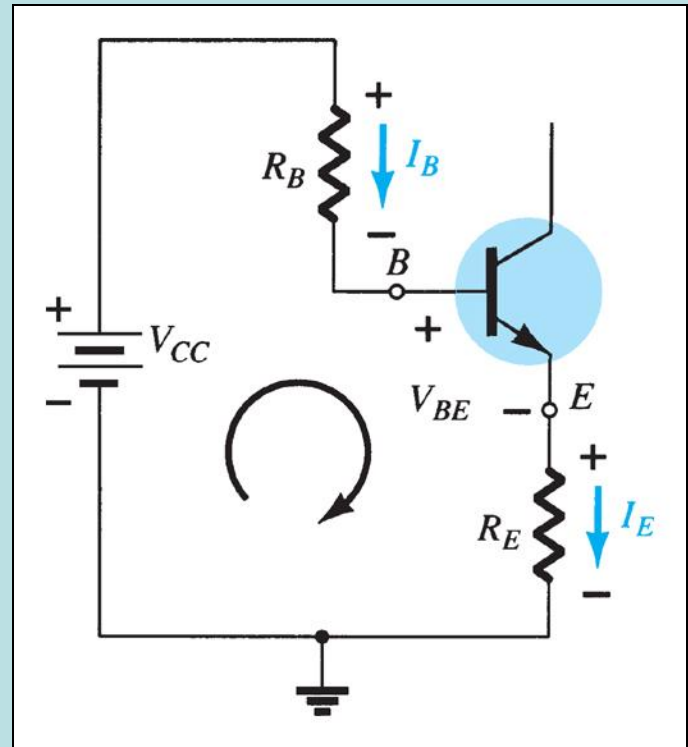
$$+V_{CC} - I_E R_E - V_{BE} - I_E R_E = 0$$

Since  $I_E = (\beta + 1)I_B$ :

$$V_{CC} - I_B R_B - (\beta + 1)I_B R_E = 0$$

Solving for  $I_B$ :

$$I_B = \frac{V_{CC} - V_{BE}}{R_B + (\beta + 1)R_E}$$



# Collector-Emitter Loop

From Kirchhoff's voltage law:

$$I_E R_E + V_{CE} + I_C R_C - V_{CC} = 0$$

Since  $I_E \cong I_C$ :

$$V_{CE} = V_{CC} - I_C (R_C + R_E)$$

Also:

$$V_E = I_E R_E$$

$$V_C = V_{CE} + V_E = V_{CC} - I_C R_C$$

$$V_B = V_{CC} - I_R R_B = V_{BE} + V_E$$

